

Toward a New Approach to Distributed Information Processing: Harnessing Group Structure

S. Sandeep Pradhan

(Joint work with Dinesh Krithivasan)

University of Michigan, Ann Arbor

Distributed Information Processing

- Proliferation of wireless sensor network applications
- Supported by distributed information processing
- Look at distributed source coding problems
- Information-theoretic perspective

Information and Coding theory: Traditional Approach

Information Theory:

- Develop efficient information processing strategies for communication
- Obtain computable performance limits
- Random coding: probability distribution on a collection of communication systems
- Show good average performance
- Encoding and decoding have exponential complexity

Information and Coding theory: Traditional Approach

Information Theory:

- Develop efficient information processing strategies for communication
- Obtain computable performance limits
- Random coding: probability distribution on a collection of communication systems
- Show good average performance
- Encoding and decoding have exponential complexity

Coding Theory:

- Approach these limits using structured codes (Ex: linear codes)
- Fast encoding and decoding algorithms
- Objective: use structured codes for practical implementability

Random Coding in multi-terminal systems

- Prob. distribution on a collection of codebooks (ensemble)
- Lot of bad codebooks in the ensemble
- Average performance significantly affected by these bad 'apples'

Random Coding in multi-terminal systems

- Prob. distribution on a collection of codebooks (ensemble)
- Lot of bad codebooks in the ensemble
- Average performance significantly affected by these bad 'apples'
- Algebraic structure can be used to weed out bad 'apples'
- Better ensemble with better performance

Random Coding in multi-terminal systems

- Prob. distribution on a collection of codebooks (ensemble)
- Lot of bad codebooks in the ensemble
- Average performance significantly affected by these bad 'apples'
- Algebraic structure can be used to weed out bad 'apples'
- Better ensemble with better performance
- Gain barely noticeable in point-to-point communication
 - Improvement in second order performance (error exponents)
 - Binary Symmetric case: almost all linear codes achieve expurgated bound without expurgation

Random Coding in multi-terminal systems

- Prob. distribution on a collection of codebooks (ensemble)
- Lot of bad codebooks in the ensemble
- Average performance significantly affected by these bad 'apples'
- Algebraic structure can be used to weed out bad 'apples'
- Better ensemble with better performance
- Gain barely noticeable in point-to-point communication
 - Improvement in second order performance (error exponents)
 - Binary Symmetric case: almost all linear codes achieve expurgated bound without expurgation
- Gains significant in multi-terminal communication

Caution: Even in point-to-point set-up

- Linear codes do not achieve in general
 - Shannon rate-distortion function
 - Shannon capacity-cost function

Caution: Even in point-to-point set-up

- Linear codes do not achieve in general
 - Shannon rate-distortion function
 - Shannon capacity-cost function
- Injection of some non-linearity appears to be necessary for optimality

Prior Work: Linear codes for multi-terminal communication

- Linear codes for symmetric source/channel coding problems
- Lattice codes for Gaussian source/channel coding problems

- Linear codes for symmetric source/channel coding problems
- Lattice codes for Gaussian source/channel coding problems
- Examples: (incomplete list)
 - Korner-Marton
 - Han-Kobayashi
 - Ahlswede-Han
 - Forney-Barg
 - Philosof-Zamir-Erez
 - Nazer-Gastpar
 - Krithivasan-Pradhan
 - Viswanath
 - ...

Contributions of the present work

- A unified approach to distributed source coding problem
- Discrete memoryless setting
- Applicable to general source statistics and distortion functions

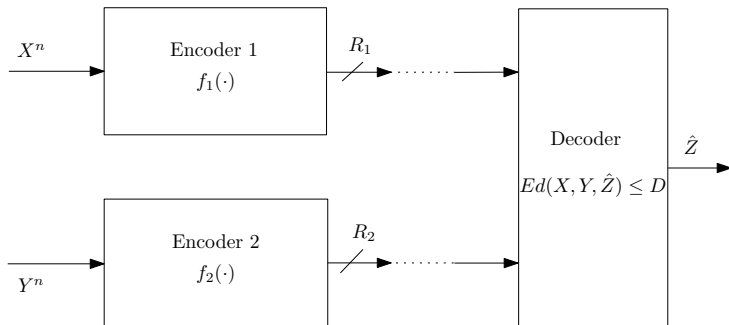
Contributions of the present work

- A unified approach to distributed source coding problem
- Discrete memoryless setting
- Applicable to general source statistics and distortion functions
- Based on abstract abelian groups: groups capture structure
- Nested codes over groups
- New rate-distortion region

Contributions of the present work

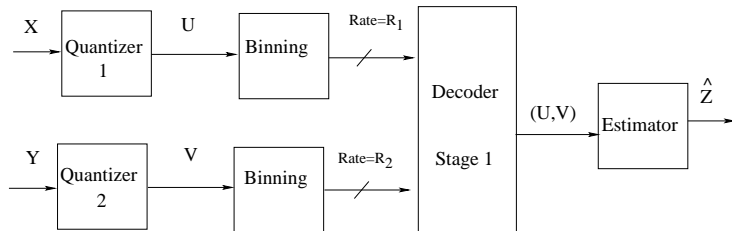
- A unified approach to distributed source coding problem
- Discrete memoryless setting
- Applicable to general source statistics and distortion functions
- Based on abstract abelian groups: groups capture structure
- Nested codes over groups
- New rate-distortion region
- Previously known rate distortion regions can be achieved using nested linear codes

A Distributed Source Coding Problem



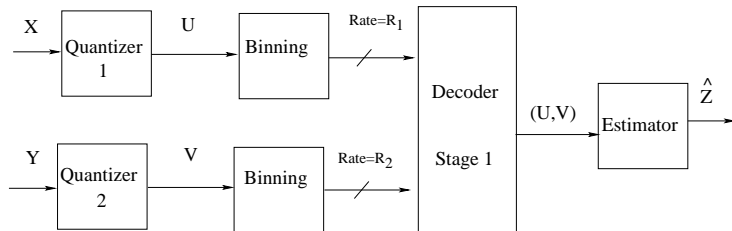
- Set of encoders observe different components of a vector source
- Central decoder receives quantized observations from the encoders
- Best known rate region - Berger-Tung based

Berger-Tung Based Rate Region: Known Results



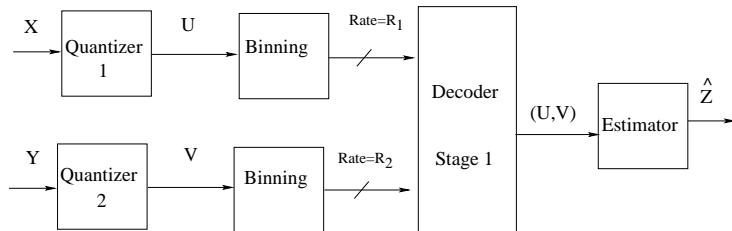
- For given distortion D
- Quantize X to U and quantize Y to V

Berger-Tung Based Rate Region: Known Results



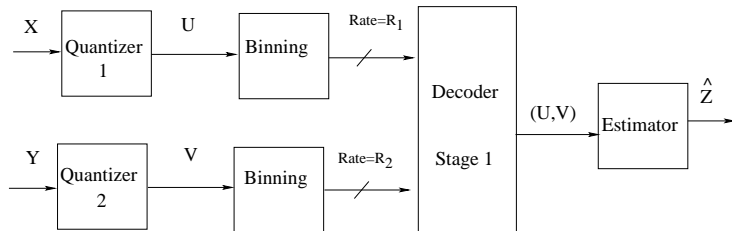
- For given distortion D
- Quantize X to U and quantize Y to V
- Find two quantizer transformations $p_{U|X}$ and $p_{V|Y}$ such that
 - best estimator $\hat{Z} = F(U, V)$ from (U, V) satisfies distortion D .

Berger-Tung Based Rate Region: Known Results



- For given distortion D
- Quantize X to U and quantize Y to V
- Find two quantizer transformations $p_{U|X}$ and $p_{V|Y}$ such that
 - best estimator $\hat{Z} = F(U, V)$ from (U, V) satisfies distortion D .
- Rates incurred in quantization = $I(X; U)$ and $I(Y; V)$

Berger-Tung Based Rate Region: Known Results



- For given distortion D
- Quantize X to U and quantize Y to V
- Find two quantizer transformations $p_{U|X}$ and $p_{V|Y}$ such that
 - best estimator $\hat{Z} = F(U, V)$ from (U, V) satisfies distortion D .
- Rates incurred in quantization = $I(X; U)$ and $I(Y; V)$
- Rate rebate by exploiting correlation between U and $V = I(U; V)$

- Hence

$$R_1 \geq I(X;U) - I(U;V)$$

$$R_2 \geq I(Y;V) - I(U;V)$$

$$R_1 + R_2 \geq I(X;U) + I(Y;V) - I(U;V)$$

- Achieved using random quantization and random binning

- Hence

$$R_1 \geq I(X; U) - I(U; V)$$

$$R_2 \geq I(Y; V) - I(U; V)$$

$$R_1 + R_2 \geq I(X; U) + I(Y; V) - I(U; V)$$

- Achieved using random quantization and random binning
- Observations:
 - Estimator $\hat{Z} = F(U, V)$ may be an information lossy transformation
 - Is it possible to reconstruct directly \hat{Z} at the decoder?
 - Can we get a rate rebate that is greater than $I(U; V)$?

- Hence

$$R_1 \geq I(X; U) - I(U; V)$$

$$R_2 \geq I(Y; V) - I(U; V)$$

$$R_1 + R_2 \geq I(X; U) + I(Y; V) - I(U; V)$$

- Achieved using random quantization and random binning
- Observations:
 - Estimator $\hat{Z} = F(U, V)$ may be an information lossy transformation
 - Is it possible to reconstruct directly \hat{Z} at the decoder?
 - Can we get a rate rebate that is greater than $I(U; V)$?
 - Can a joint design of quantizer and binning get better performance?

Example 1

- X, Y - 3 bit correlated binary sources, $d_H(X, Y) \leq 1$
- Decoder interested in reconstructing
$$\hat{Z} = X \oplus_2 Y \in \{000, 001, 010, 100\}$$

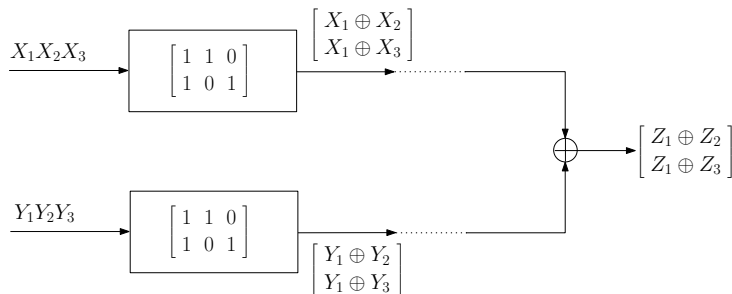
Example 1

- X, Y - 3 bit correlated binary sources, $d_H(X, Y) \leq 1$
- Decoder interested in reconstructing
$$\hat{Z} = X \oplus_2 Y \in \{000, 001, 010, 100\}$$
- Berger-Tung based coding scheme:
 - Reconstruct sources X, Y . Compute $\hat{Z} = X \oplus_2 Y$
 - Sum rate: $H(X, Y) = 5$ bits

Example 1

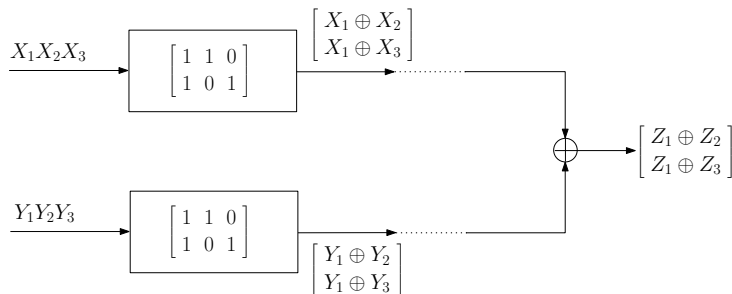
- X, Y - 3 bit correlated binary sources, $d_H(X, Y) \leq 1$
- Decoder interested in reconstructing
$$\hat{Z} = X \oplus_2 Y \in \{000, 001, 010, 100\}$$
- Berger-Tung based coding scheme:
 - Reconstruct sources X, Y . Compute $\hat{Z} = X \oplus_2 Y$
 - Sum rate: $H(X, Y) = 5$ bits
- Can we do better?

Example 1: Linear Coding Scheme



- $X_1 \oplus X_2 \oplus Y_1 \oplus Y_2 = X_1 \oplus Y_1 \oplus X_2 \oplus Y_2 = \hat{Z}_1 \oplus \hat{Z}_2$
- Sum rate: $2 + 2 = 4$ bits

Example 1: Linear Coding Scheme



- $X_1 \oplus X_2 \oplus Y_1 \oplus Y_2 = X_1 \oplus Y_1 \oplus X_2 \oplus Y_2 = \hat{Z}_1 \oplus \hat{Z}_2$
- Sum rate: $2 + 2 = 4$ bits
- Significant features:
 - encoding function **commutes** with function $\hat{Z} = X \oplus Y$
 - Identical binning at both encoders
 - Linear codes

Example 2: Reconstruct the pair (X, Y) (Slepian-Wolf)

- (X, Y) - binary correlated sources

Example 2: Reconstruct the pair (X, Y) (Slepian-Wolf)

- (X, Y) - binary correlated sources
- Can be thought of as addition in \mathbb{F}_4

	00	01	10	11
00	00	01	10	11
01	01	00	11	10
10	10	11	00	01
11	11	10	01	00

- Map binary sources into \mathbb{F}_4

Example 2: Reconstruct the pair (X, Y) (Slepian-Wolf)

- (X, Y) - binary correlated sources
- Can be thought of as addition in \mathbb{F}_4

	00	01	10	11
00	00	01	10	11
01	01	00	11	10
10	10	11	00	01
11	11	10	01	00

- Map binary sources into \mathbb{F}_4
- Encode sequentially one digit at a time
- Previously decoded digits = side information at the decoder

Example 3: Reconstruct the function $X \vee Y$

- (X, Y) - binary correlated sources

Example 3: Reconstruct the function $X \vee Y$

- (X, Y) - binary correlated sources
- Can be embedded in the addition table in \mathbb{F}_3

	0	1	2
0	0	1	2
1	1	2	0
2	2	0	1

- Map binary sources into \mathbb{F}_3

Example 3: Reconstruct the function $X \vee Y$

- (X, Y) - binary correlated sources
- Can be embedded in the addition table in \mathbb{F}_3

	0	1	2
0	0	1	2
1	1	2	0
2	2	0	1

- Map binary sources into \mathbb{F}_3
- Construct linear codes over \mathbb{F}_3
- Can do better than Slepian-Wolf coding

Overview of our Coding Scheme

- Fix test channel $p_{U|X}, p_{V|Y}$

Overview of our Coding Scheme

- Fix test channel $p_{U|X}, p_{V|Y}$
- Function to be reconstructed $F(U, V)$ - embed in the addition table of some abelian group
- Abelian groups decomposable into primary cyclic groups

Overview of our Coding Scheme

- Fix test channel $p_{U|X}, p_{V|Y}$
- Function to be reconstructed $F(U, V)$ - embed in the addition table of some abelian group
- Abelian groups decomposable into primary cyclic groups
- Encode sequentially using **nested group codes**
- All codes used in encoding have the same algebraic structure

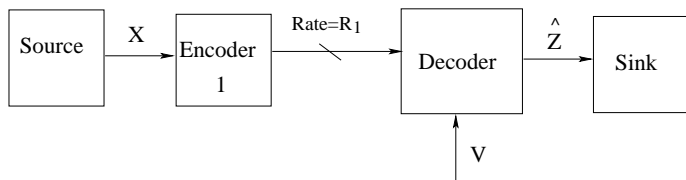
Overview of our Coding Scheme

- Fix test channel $p_{U|X}, p_{V|Y}$
- Function to be reconstructed $F(U, V)$ - embed in the addition table of some abelian group
- Abelian groups decomposable into primary cyclic groups
- Encode sequentially using **nested group codes**
- All codes used in encoding have the same algebraic structure
- Two quantizers and the binning operation is designed jointly
- A framework applicable to arbitrary source statistics and distortion measures

Overview of our Coding Scheme

- Fix test channel $p_{U|X}, p_{V|Y}$
- Function to be reconstructed $F(U, V)$ - embed in the addition table of some abelian group
- Abelian groups decomposable into primary cyclic groups
- Encode sequentially using **nested group codes**
- All codes used in encoding have the same algebraic structure
- Two quantizers and the binning operation is designed jointly
- A framework applicable to arbitrary source statistics and distortion measures
- We looked at linear codes for binning.
- How to do quantization using structured codes?

Berger-Tung Rate Region: Closer Look

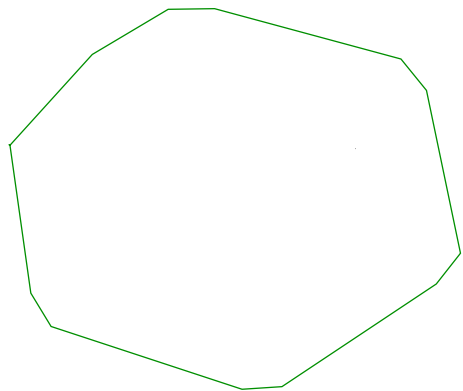


- $R_2 = I(Y; V)$, $R_1 = I(X; U) - I(U; V)$
- Source X : alphabet \mathcal{X} , distribution p_X
- Side Information V : alphabet \mathcal{V} , distribution $p_{V|X}$.
- Reconstruction: alphabet $\hat{\mathcal{Z}}$.
- Compress X into bits to achieve a target distortion.

Encoding: Quantization + Binning

- Quantize X to U with rate $I(X; U)$
- Partition the quantizer into bins of rate $I(U; V)$
- Each bin is a good channel code for the channel $p_{V|U}$.
- Send the bin index to the decoder
- Recover the quantizer codeword from the bin using V .

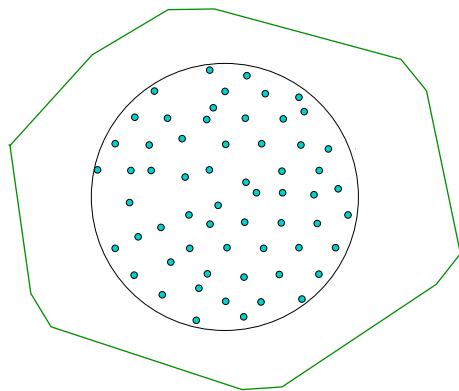
Space in which quantizer is built



- Space: U^n
- Quantizer : X to U

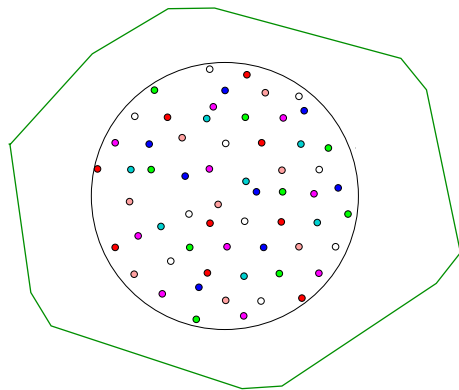
- Joint histogram of source word and its quantized version $\approx p_{XU}$

Good Quantizer



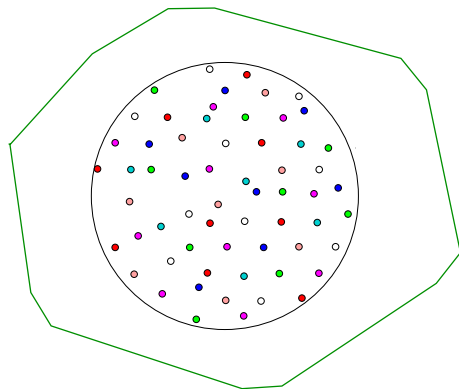
- Quantize X to U
- Must cover a specific region
- Typical set with respect to p_U .
- Rate: $I(X; U)$.
- Shannon source code

Quantizer Partition into bins

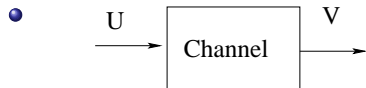


- Partition into bins

Quantizer Partition into bins

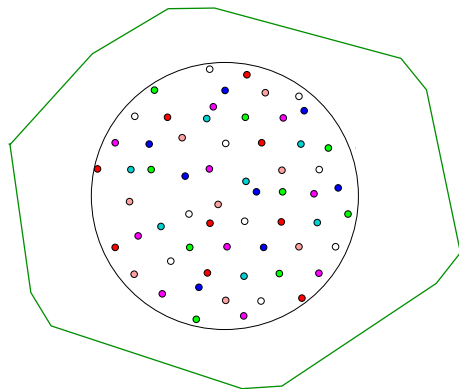


- Partition into bins
- Bin = Good channel code
- Fictitious channel

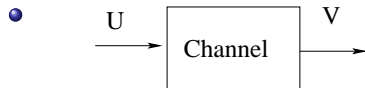


- I/P Distribution: P_U
- Conditional distribution: $p_{V|U}$.

Quantizer Partition into bins

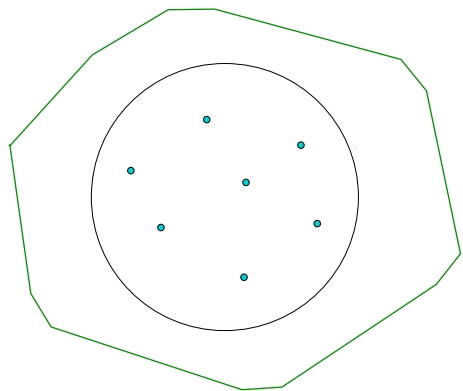


- Partition into bins
- Bin = Good channel code
- Fictitious channel



- I/P Distribution: P_U
- Conditional distribution: $p_{V|U}$.
- Bin Rate: $I(U; V)$.
- Bin density rate:
$$= I(X; U) - I(U; V)$$

Good Channel Code



- Each bin is a good channel code
- Pack codewords in the region
- Typical set with respect to U
- Rate: $I(U; V)$.
- Shannon channel code

Illustration of Encoding

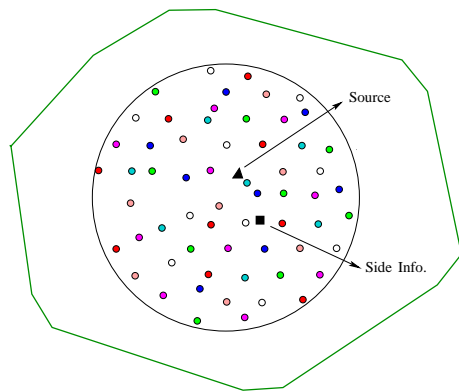


Illustration of Decoding

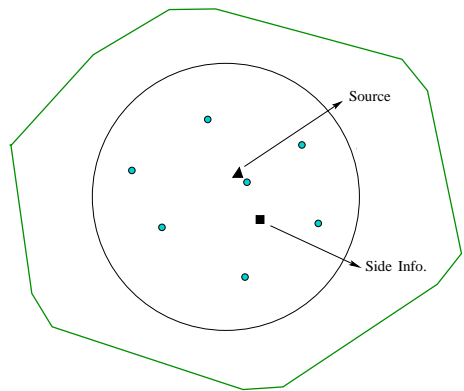
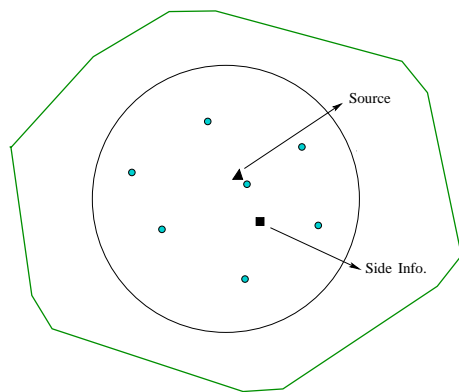


Illustration of Decoding



- None of the codes used here have any algebraic structure

RECALL

- Linear Codes do not achieve
 - Shannon rate-distortion function
 - Shannon capacity-cost function
- Linear code do not achieve
 - the source rate $I(X; U)$
 - the channel rate $I(U; V)$

Try this with linear codes

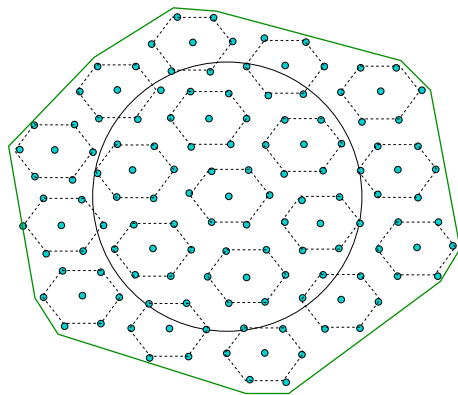
RECALL

- Linear Codes do not achieve
 - Shannon rate-distortion function
 - Shannon capacity-cost function
- Linear code do not achieve
 - the source rate $I(X; U)$
 - the channel rate $I(U; V)$

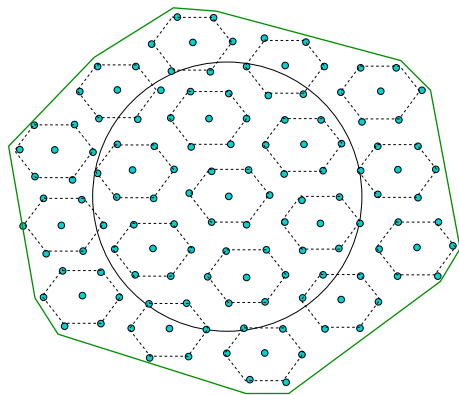
Q: How to achieve $I(X; U) - I(U; V)$ using linear codes?

Linear code as a Quantizer: Lemma 1

- Linear code \mathcal{C}_1

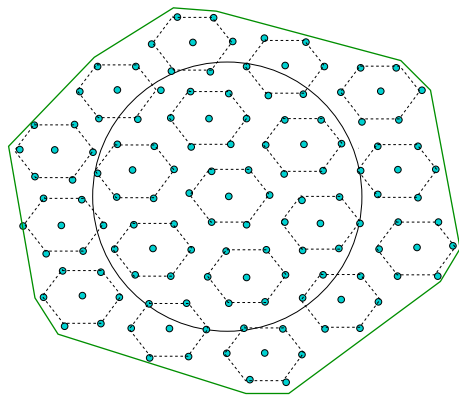


Linear code as a Quantizer: Lemma 1



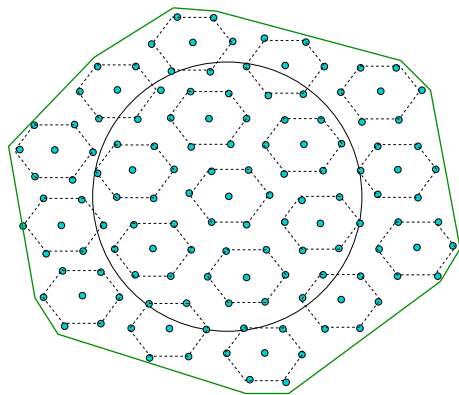
- Linear code \mathcal{C}_1
- Contains a good quantizer
- True for arbitrary $(\mathcal{X}, \mathcal{U}, p_{XU})$

Linear code as a Quantizer: Lemma 1



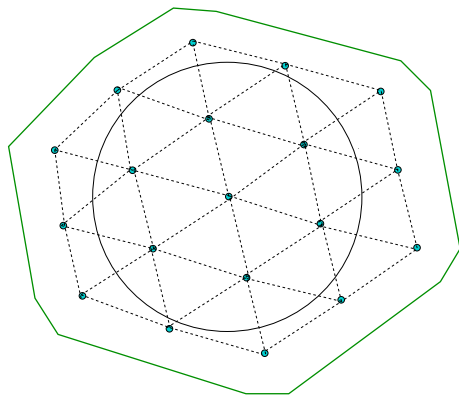
- Linear code \mathcal{C}_1
- Contains a good quantizer
- True for arbitrary $(\mathcal{X}, \mathcal{U}, p_{XU})$
- Codeword density as before
- Expand beyond typical set

Linear code as a Quantizer: Lemma 1



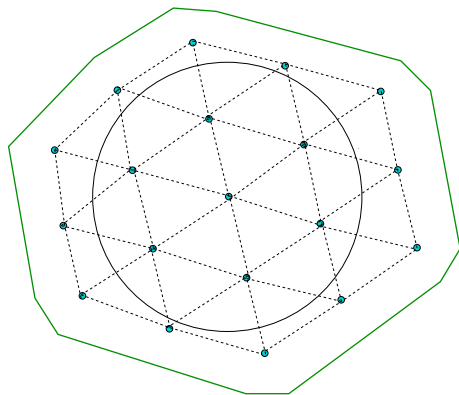
- Linear code \mathcal{C}_1
- Contains a good quantizer
- True for arbitrary $(\mathcal{X}, \mathcal{U}, p_{XU})$
- Codeword density as before
- Expand beyond typical set
- Rate of the code:
 $\log |\mathcal{U}| - H(U|X)$
- Penalty for linearity:
 $\log |\mathcal{U}| - H(U)$.
- Refer: Good linear source code

Linear code as a channel code: Lemma 2



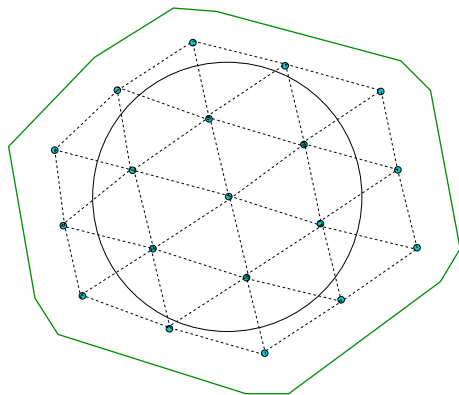
- Linear code \mathcal{C}_2
- Every coset contains a good channel code

Linear code as a channel code: Lemma 2



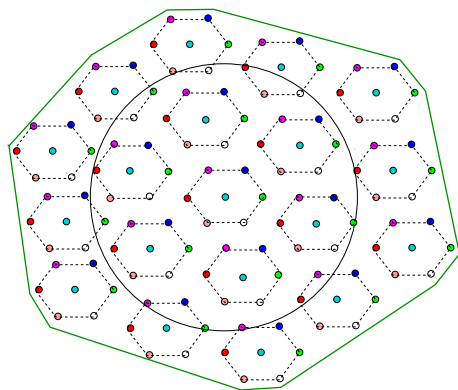
- Linear code \mathcal{C}_2
- Every coset contains a good channel code
- True for arbitrary $(\mathcal{U}, \mathcal{V}, p_{UV})$
- Codeword density as before
- Expand beyond typical set

Linear code as a channel code: Lemma 2



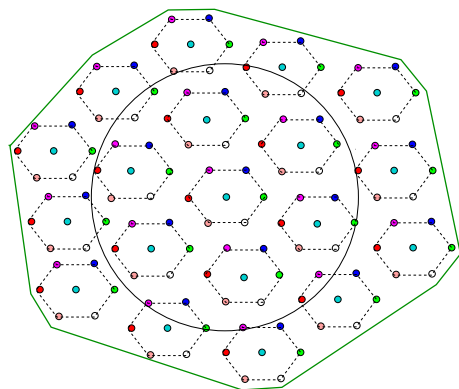
- Linear code \mathcal{C}_2
- Every coset contains a good channel code
- True for arbitrary $(\mathcal{U}, \mathcal{V}, p_{UV})$
- Codeword density as before
- Expand beyond typical set
- Rate of the code:
 $\log |\mathcal{U}| - H(U|V)$
- Penalty for linearity:
 $\log |\mathcal{U}| - H(U)$
- Refer: Good linear channel code

Linear code Partition



- Partition \mathcal{C}_1 into cosets of \mathcal{C}_2
- Coset density as before

Linear code Partition



- Partition \mathcal{C}_1 into cosets of \mathcal{C}_2
- Coset density as before

- Coset density rate
$$= \log |\mathcal{U}| - H(U|X) - \log |\mathcal{U}| + H(U|V)$$
$$= I(X;U) - I(U;V)$$

- Built on Galois fields
- Nested linear codes can achieve Berger-Tung bound
- Good nested linear codes can achieve Shannon limit
 - Take $V = \text{constant}$
 - Source Code Rate: $\log |\mathcal{U}| - H(U|X)$
 - Channel Code Rate: $\log |\mathcal{U}| - H(U)$
- A specific form of non-linearity

Linear Codes: Upshot

- Built on Galois fields
- Nested linear codes can achieve Berger-Tung bound
- Good nested linear codes can achieve Shannon limit
 - Take $V = \text{constant}$
 - Source Code Rate: $\log |\mathcal{U}| - H(U|X)$
 - Channel Code Rate: $\log |\mathcal{U}| - H(U)$
- A specific form of non-linearity

Next: extension to arbitrary abelian groups

Groups - An Introduction

- G - a finite abelian group of order n
- $G \cong \mathbb{Z}_{p_1^{e_1}} \times \mathbb{Z}_{p_2^{e_2}} \cdots \times \mathbb{Z}_{p_k^{e_k}}$
- G isomorphic to direct product of possibly repeating primary cyclic groups

$$g \in G \Leftrightarrow g = (g_1, \dots, g_k), \quad g_i \in \mathbb{Z}_{p_i^{e_i}}$$

- Call g_i as the i th digit of g

Groups - An Introduction

- G - a finite abelian group of order n
- $G \cong \mathbb{Z}_{p_1^{e_1}} \times \mathbb{Z}_{p_2^{e_2}} \cdots \times \mathbb{Z}_{p_k^{e_k}}$
- G isomorphic to direct product of possibly repeating primary cyclic groups

$$g \in G \Leftrightarrow g = (g_1, \dots, g_k), \quad g_i \in \mathbb{Z}_{p_i^{e_i}}$$

- Call g_i as the i th digit of g
- Enough to prove coding theorems for primary cyclic groups
- Extension to arbitrary abelian groups through digit decomposition

Example 4:

- Let group size be 36
- $36 = 2^2 \times 3^2$
- Abelian groups of order 36:
 - $\mathbb{Z}_4 \times \mathbb{Z}_9$: Two digits
 - $\mathbb{Z}_2 \times \mathbb{Z}_2 \times \mathbb{Z}_9$: Three digits
 - $\mathbb{Z}_4 \times \mathbb{Z}_3 \times \mathbb{Z}_3$: Three digits
 - $\mathbb{Z}_2 \times \mathbb{Z}_2 \times \mathbb{Z}_3 \times \mathbb{Z}_3$: Four digits

Embedding a function in a group G

A function $F : \mathcal{U} \times \mathcal{V} \rightarrow \hat{\mathcal{Z}}$ can be embedded in G if

Embedding a function in a group G

A function $F : \mathcal{U} \times \mathcal{V} \rightarrow \hat{\mathcal{Z}}$ can be embedded in G if

- \exists a one-to-one mapping $F_1 : \mathcal{U} \rightarrow G$
- \exists a one-to-one mapping $F_2 : \mathcal{V} \rightarrow G$
- \exists a mapping $F_3 : G \rightarrow \hat{\mathcal{Z}}$
- such that $F(U, V) = F_3 [F_1(U) \oplus_G F_2(V)]$

Embedding a function in a group G

A function $F : \mathcal{U} \times \mathcal{V} \rightarrow \hat{\mathcal{Z}}$ can be embedded in G if

- \exists a one-to-one mapping $F_1 : \mathcal{U} \rightarrow G$
- \exists a one-to-one mapping $F_2 : \mathcal{V} \rightarrow G$
- \exists a mapping $F_3 : G \rightarrow \hat{\mathcal{Z}}$
- such that $F(U, V) = F_3 [F_1(U) \oplus_G F_2(V)]$

Example: $\hat{\mathcal{Z}} = U \vee V$ can be embedded in \mathbb{Z}_3

Nested Group Codes - Motivation

- Codes used in KM,SW - good channel codes
 - Cosets bin the entire space
 - Suitable for lossless coding

Nested Group Codes - Motivation

- Codes used in KM,SW - good channel codes
 - Cosets bin the entire space
 - Suitable for lossless coding
- Lossy coding: Need to quantize first
 - Dilute coset density - Nested group codes
 - Fine code - Quantizes the sources
 - Coarse code - Bins only the fine code

Nested Group Codes

- Group code over $\mathbb{Z}_{p^r}^n$: $\mathcal{C} < \mathbb{Z}_{p^r}^n$
- $\mathcal{C} = \ker(\phi)$ for some homomorphism $\phi: \mathbb{Z}_{p^r}^n \rightarrow \mathbb{Z}_{p^r}^k$

Nested Group Codes

- Group code over $\mathbb{Z}_{p^r}^n$: $\mathcal{C} < \mathbb{Z}_{p^r}^n$
- $\mathcal{C} = \ker(\phi)$ for some homomorphism $\phi: \mathbb{Z}_{p^r}^n \rightarrow \mathbb{Z}_{p^r}^k$
- $(\mathcal{C}_1, \mathcal{C}_2)$ nested if $\mathcal{C}_2 \subset \mathcal{C}_1$

Nested Group Codes

- Group code over $\mathbb{Z}_{p^r}^n$: $\mathcal{C} < \mathbb{Z}_{p^r}^n$
- $\mathcal{C} = \ker(\phi)$ for some homomorphism $\phi: \mathbb{Z}_{p^r}^n \rightarrow \mathbb{Z}_{p^r}^k$
- $(\mathcal{C}_1, \mathcal{C}_2)$ nested if $\mathcal{C}_2 \subset \mathcal{C}_1$
- We need:
 - $\mathcal{C}_1 < \mathbb{Z}_{p^r}^n$: “good” source code
 - Can find $u^n \in \mathcal{C}_1$ jointly typical with source x^n
 - $\mathcal{C}_2 < \mathbb{Z}_{p^r}^n$: “good” channel code
 - Can distinguish between typical channel noise sequences

Good Group Source Codes

- Good group source code \mathcal{C}_1 for the triple $(\mathcal{X}, \mathcal{U}, P_{XU})$
- Assume $\mathcal{U} = \mathbb{Z}_{p^r}$ for some prime p and exponent $r > 0$

Good Group Source Codes

- Good group source code \mathcal{C}_1 for the triple $(\mathcal{X}, \mathcal{U}, P_{XU})$
- Assume $\mathcal{U} = \mathbb{Z}_{p^r}$ for some prime p and exponent $r > 0$

Lemma

Exists for large n if

$$\frac{1}{n} \log |\mathcal{C}_1| \geq \log p^r - \min\{H(U|X), r|H(U|X) - \log p^{r-1}|^+\}$$

Good Group Source Codes

- Good group source code \mathcal{C}_1 for the triple $(\mathcal{X}, \mathcal{U}, P_{XU})$
- Assume $\mathcal{U} = \mathbb{Z}_{p^r}$ for some prime p and exponent $r > 0$

Lemma

Exists for large n if

$$\frac{1}{n} \log |\mathcal{C}_1| \geq \log p^r - \min\{H(U|X), r|H(U|X) - \log p^{r-1}|^+\}$$

- Compare with optimal random code: $H(U) - H(U|X) = I(X; U)$
- Compare with linear code: $\log p^r - H(U|X)$
- Not good in Shannon sense
- Extra penalty for imposing group structure beyond linearity

Good Group Channel Codes

- Good group channel code \mathcal{C}_2 for the triple $(\mathcal{U}, \mathcal{V}, P_{UV})$
- Assume $\mathcal{U} = \mathbb{Z}_{p^r}$ for some prime p and exponent $r > 0$

Good Group Channel Codes

- Good group channel code \mathcal{C}_2 for the triple $(\mathcal{U}, \mathcal{V}, P_{UV})$
- Assume $\mathcal{U} = \mathbb{Z}_{p^r}$ for some prime p and exponent $r > 0$

Lemma

Exists for large n if

$$\frac{1}{n} \log |\mathcal{C}_2| \leq \log p^r - \max_{0 \leq i < r} \binom{r}{r-i} (H(U|V) - H([U]_i|V))$$

Good Group Channel Codes

- Good group channel code \mathcal{C}_2 for the triple $(\mathcal{U}, \mathcal{V}, P_{UV})$
- Assume $\mathcal{U} = \mathbb{Z}_{p^r}$ for some prime p and exponent $r > 0$

Lemma

Exists for large n if

$$\frac{1}{n} \log |\mathcal{C}_2| \leq \log p^r - \max_{0 \leq i < r} \left(\frac{r}{r-i} \right) (H(U|V) - H([U]_i|V))$$

- $[U]_i$ is a function of U .
- Compare with optimal random code: $H(U) - H(U|V)$
- Compare with linear code: $\log p^r - H(U|V)$
- Not good in Shannon sense
- Extra penalty for imposing group structure beyond linearity

- Matrix characterization of subgroups of direct product of a group
- random coding over subgroups
- Suen's inequality [1998]

Coding Approach: 5 steps

Fix $p_{U|X}, p_{V|Y}$ such that $\mathbb{E}d(X, Y, F(U, V)) \leq D$

Coding Approach: 5 steps

Fix $p_{U|X}, p_{V|Y}$ such that $\mathbb{E}d(X, Y, F(U, V)) \leq D$

- Step 1: Embed $F(U, V)$ in an abelian group G
- Step 2: Decompose G into primary cyclic groups: G_1, \dots, G_K
 - Represent $U = (U_1, \dots, U_K)$ and $V = (V_1, \dots, V_K)$

Coding Approach: 5 steps

Fix $p_{U|X}, p_{V|Y}$ such that $\mathbb{E}d(X, Y, F(U, V)) \leq D$

- Step 1: Embed $F(U, V)$ in an abelian group G
- Step 2: Decompose G into primary cyclic groups: G_1, \dots, G_K
 - Represent $U = (U_1, \dots, U_K)$ and $V = (V_1, \dots, V_K)$
- Step 3: Group source code over G_i for every i : $\mathcal{C}_{11}(i), \mathcal{C}_{12}(i)$
- Step 4: Group channel code over G_i for every i : $\mathcal{C}_2(i)$

Coding Approach: 5 steps

Fix $p_{U|X}, p_{V|Y}$ such that $\mathbb{E}d(X, Y, F(U, V)) \leq D$

- Step 1: Embed $F(U, V)$ in an abelian group G
- Step 2: Decompose G into primary cyclic groups: G_1, \dots, G_K
 - Represent $U = (U_1, \dots, U_K)$ and $V = (V_1, \dots, V_K)$
- Step 3: Group source code over G_i for every i : $\mathcal{C}_{11}(i), \mathcal{C}_{12}(i)$
- Step 4: Group channel code over G_i for every i : $\mathcal{C}_2(i)$
- Step 5: Nest the channel code inside the source codes
 - $\mathcal{C}_2(i) < \mathcal{C}_{11}(i)$ and $\mathcal{C}_2(i) < \mathcal{C}_{12}(i)$
 - Identical binning of quantizers

Encoding and Decoding

Encoders: at the i th stage

- Encode the sources X and Y to digits U_i and V_i sequentially
 - quantize + bin

Encoding and Decoding

Encoders: at the i th stage

- Encode the sources X and Y to digits U_i and V_i sequentially
 - quantize + bin

Decoder: at the i th stage

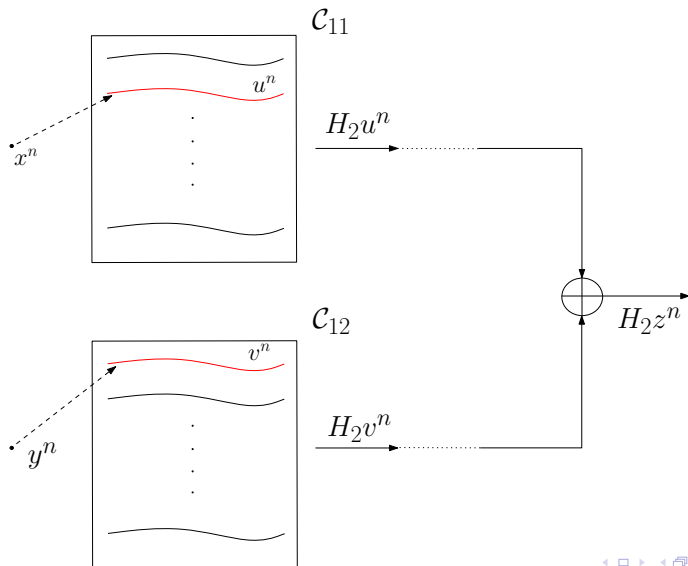
- Recover $\hat{Z}_i = U_i \oplus_{G_i} V_i$
- Use previously decoded digits as side information

Example 5

Suppose we embed $F(U, V)$ in $\mathbb{Z}_4 \times \mathbb{Z}_7$

- We have two digits: (U_1, V_1, \hat{Z}_1) and (U_2, V_2, \hat{Z}_2)
- Two stages
- Stage 1: \mathbb{Z}_4 operation
- Stage 2: \mathbb{Z}_7 operation

Coding Strategy: Nested group codes $\mathcal{C}_2 < \mathcal{C}_{11}, \mathcal{C}_{12}$



Theorem

The set of tuples (R_1, R_2, D) that satisfy

$$R_1 \geq \max_{0 \leq i < r} \left(\frac{r}{r-i} \right) (H(\hat{Z}) - H([\hat{Z}]_i)) - \min\{H(U|X), r|H(U|X) - \log p^{r-1}|^+\}$$

$$R_2 \geq \max_{0 \leq i < r} \left(\frac{r}{r-i} \right) (H(\hat{Z}) - H([\hat{Z}]_i)) - \min\{H(V|Y), r|H(V|Y) - \log p^{r-1}|^+\}$$

$$D \geq \mathbb{E}d(X, Y, F(U, V))$$

are achievable.

Theorem

The set of tuples (R_1, R_2, D) that satisfy

$$R_1 \geq \max_{0 \leq i < r} \left(\frac{r}{r-i} \right) (H(\hat{Z}) - H([\hat{Z}]_i)) - \min\{H(U|X), r|H(U|X) - \log p^{r-1}|^+\}$$

$$R_2 \geq \max_{0 \leq i < r} \left(\frac{r}{r-i} \right) (H(\hat{Z}) - H([\hat{Z}]_i)) - \min\{H(V|Y), r|H(V|Y) - \log p^{r-1}|^+\}$$

$$D \geq \mathbb{E}d(X, Y, F(U, V))$$

are achievable.

- More general rate region possible by
 - Embedding in general groups and using digit decomposition
 - Alternative coding strategy at i th stage - Encode (U_i, V_i) instead of \hat{Z}_i

Example 6: Lossless reconstruction of quaternary function

- X and Y take values in $\{0, 1, 2, 3\}$
- Lossless reconstruction of $\hat{Z} = X - Y \bmod 4$

Example 6: Lossless reconstruction of quaternary function

- X and Y take values in $\{0, 1, 2, 3\}$
- Lossless reconstruction of $\hat{Z} = X - Y \bmod 4$
- Can be embedded in abelian groups of order ≤ 16
- \mathbb{Z}_4 , \mathbb{Z}_7 , $\mathbb{Z}_2 \oplus \mathbb{Z}_2 \oplus \mathbb{Z}_2$, $\mathbb{Z}_4 \oplus \mathbb{Z}_4$ give non-trivial performance

Example 6: Lossless reconstruction of quaternary function

- X and Y take values in $\{0, 1, 2, 3\}$
- Lossless reconstruction of $\hat{Z} = X - Y \bmod 4$
- Can be embedded in abelian groups of order ≤ 16
- \mathbb{Z}_4 , \mathbb{Z}_7 , $\mathbb{Z}_2 \oplus \mathbb{Z}_2 \oplus \mathbb{Z}_2$, $\mathbb{Z}_4 \oplus \mathbb{Z}_4$ give non-trivial performance

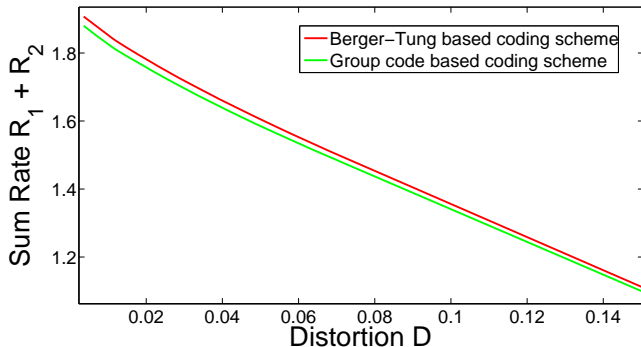
p_X	p_E	$R_{\mathbb{Z}_4}$	$R_{\mathbb{Z}_7}$	$R_{\mathbb{Z}_2 \oplus \mathbb{Z}_2 \oplus \mathbb{Z}_2}$	$R_{\mathbb{Z}_4 \oplus \mathbb{Z}_4}$
$[\frac{1}{4} \frac{1}{4} \frac{1}{4} \frac{1}{4}]$	$[\frac{1}{2} 0 \frac{1}{4} \frac{1}{4}]$	3	3.9056	3.1887	3.5
$[\frac{3}{10} \frac{6}{10} \frac{1}{10} 0]$	$[0 \frac{4}{5} \frac{1}{20} \frac{3}{20}]$	2.3911	2.0797	2.4529	2.1796
$[\frac{1}{3} \frac{1}{10} \frac{1}{2} \frac{1}{15}]$	$[\frac{3}{7} \frac{1}{7} \frac{1}{7} \frac{2}{7}]$	3.6847	4.5925	3.3495	3.4633
$[\frac{9}{10} \frac{1}{30} \frac{1}{30} \frac{1}{30}]$	$[\frac{3}{20} \frac{3}{4} \frac{1}{20} \frac{1}{20}]$	2.308	2.7065	1.9395	1.7815

Table: Example distributions for which embedding in a given group gives the lowest sum rate.

Example 7: Lossy Reconstruction of binary XOR

- Correlated binary sources (X, Y)
- Reconstruct $\hat{Z} = X \oplus_2 Y$ within Hamming distortion D
- U, V - binary auxiliary random variables
- $F(U, V)$ - one of 16 possibilities depending on $(p_{U|X}, p_{V|Y})$

Comparison of the two lower convex envelopes



- Rate gains over the Berger-Tung based scheme
- Implies Berger-Tung inner bound not tight for three-user case

- Lossless compression using group codes - achievable rates
- Lossy compression for arbitrary sources and distortion measures using group codes
- Nested linear codes - Shannon rate-distortion bound for arbitrary sources and additive distortion measures

- Lossless compression using group codes - achievable rates
- Lossy compression for arbitrary sources and distortion measures using group codes
- Nested linear codes - Shannon rate-distortion bound for arbitrary sources and additive distortion measures
- Recovers known rate regions (using nested linear codes) of
 - Berger-Tung problem
 - Wyner-Ziv problem, Wyner-Ahlsvede-Korner problem
 - Yeung-Berger problem
 - Slepian-Wolf problem, Korner-Marton problem

- Presented a nested group codes based coding scheme
- Recovered known rate regions of several distributed source coding problems
- Offers rate gains over the Berger-Tung based coding scheme
- Extensions:
 - Codes over groups for multi-user channel coding problems
 - Codes over non-abelian groups