

SIGNAL SUBSPACES, ORTHOGONAL PROJECTIONS, AND LEAST SQUARES ESTIMATION

The Signal Subspace Model

Let $\underline{a}_1, \dots, \underline{a}_p \in \mathbb{R}^N$ (or \mathbb{C}^N) be linearly independent (so $p \leq N$), and consider the $N \times p$ matrix

$$A = [\underline{a}_1 \ \dots \ \underline{a}_p].$$

Let $\langle A \rangle$ denote the linear span of the columns of A (equivalently, the image of A). Then

$$\dim(\langle A \rangle) = \text{rank}(A) = p$$

$$\dim(\langle A \rangle^\perp) = N - p$$

Let $\underline{b}_1, \dots, \underline{b}_{N-p} \in \mathbb{R}^N$ (or \mathbb{C}^N) be a basis for $\langle A \rangle^\perp$ and set

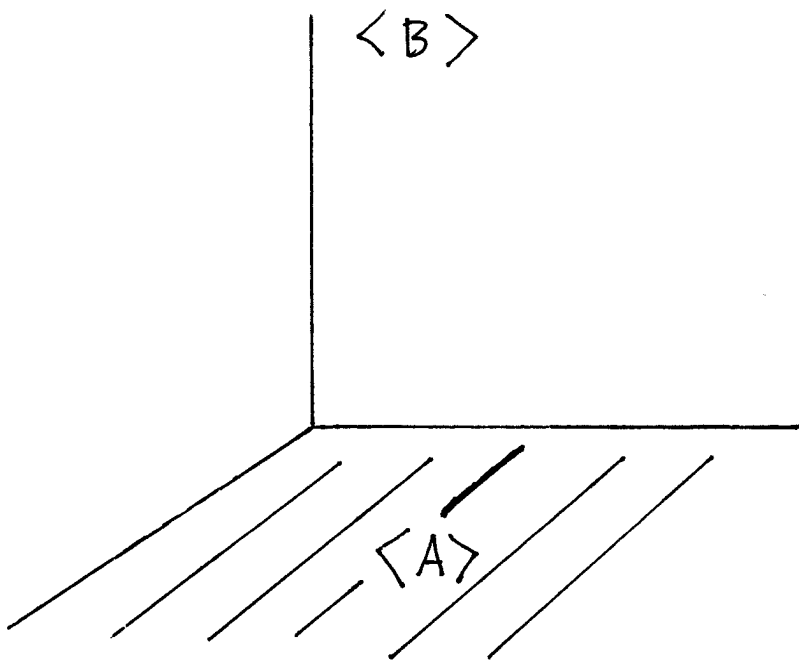
$$B = [\underline{b}_1 \ \dots \ \underline{b}_{N-p}] \quad N \times (N-p)$$

Then

$$\langle \underline{a}_i, \underline{b}_j \rangle = 0 \quad \begin{array}{l} i = 1, \dots, p \\ j = 1, \dots, N-p. \end{array}$$

(a) and $\{\underline{a}_1, \dots, \underline{a}_p, \underline{b}_1, \dots, \underline{b}_{N-p}\}$ is a _____.

The subspaces $\langle A \rangle$ and $\langle B \rangle$ form an orthogonal decomposition of \mathbb{R}^N (or \mathbb{C}^N)



$$\langle A \rangle \oplus \langle B \rangle = \mathbb{R}^N \quad (\text{or } \mathbb{C}^N)$$

Note | The vectors $\underline{a}_1, \dots, \underline{a}_p$ are not necessarily orthogonal among themselves. The same goes for $\underline{b}_1, \dots, \underline{b}_{N-p}$.

In the signal subspace model, we assume our observed signal \underline{x} has the form

$$\underline{x} = \underline{s} + \underline{w}$$

where

$\underline{s} \in \langle A \rangle$ is the signal of interest

\underline{w} is entirely noise.

We use the following terminology:

$$\langle A \rangle =$$

$$\langle B \rangle =$$

even though $w \notin \langle B \rangle$ in general.

(b)

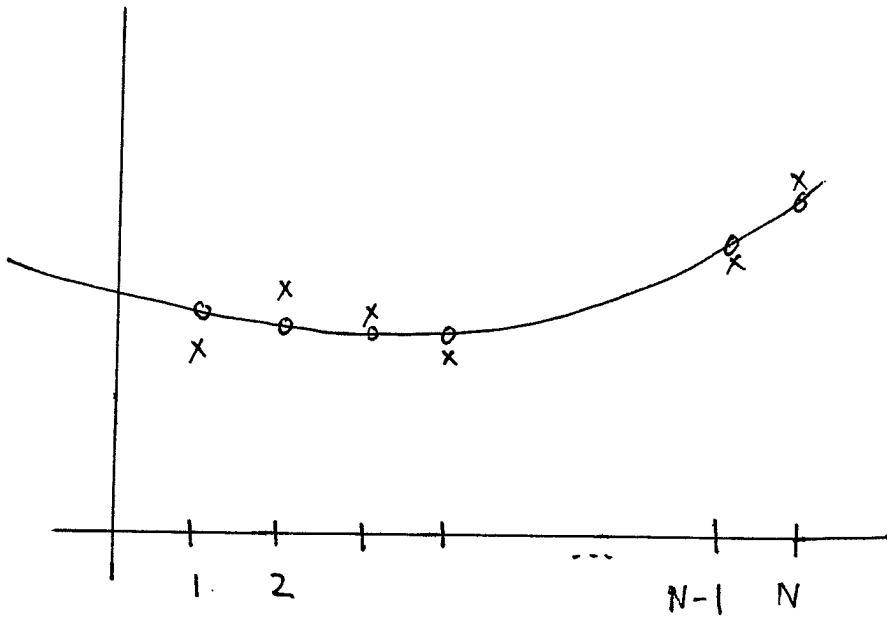
Example] Quadratic polynomial model

$$s(n) = \theta_2 n^2 + \theta_1 n + \theta_0, \quad n = 1, \dots, N$$

$$\Rightarrow \underline{s} = A \underline{\theta} \quad \text{where}$$

$$A = \begin{bmatrix} 1 & 1 & 1 \\ 1 & 2 & 4 \\ \vdots & \vdots & \vdots \\ 1 & N & N^2 \end{bmatrix}$$

$$\underline{\theta} = \begin{bmatrix} \theta_0 \\ \theta_1 \\ \theta_2 \end{bmatrix} \in \mathbb{R}^3$$



- o \rightarrow clean signal (unobserved)
- x \rightarrow noisy signal (observed)

Exercise | Consider the sinusoidal signal

$$s(n) = D \cdot \cos(2\pi fn + \phi), \quad n=0, \dots, N-1$$

where f is known but D, ϕ are unknown.

Express $\underline{s} = [s(0) \dots s(N-1)]^T$ as an element in a two-dimensional subspace.

That is, write

$$\underline{s} = A \cdot \underline{\theta}$$

where A is a known $N \times 2$ matrix and $\underline{\theta}$ is unknown.

Solution 1 Use $\cos(\alpha) = \frac{e^{j\alpha} + e^{-j\alpha}}{2}$. Then

$$s(n) = \underbrace{\left(\frac{D}{2} e^{j\phi}\right)}_{\theta_1} e^{2\pi jfn} + \underbrace{\left(\frac{D}{2} e^{-j\phi}\right)}_{\theta_2} e^{-2\pi jfn}$$

$$\Rightarrow \underline{s} = \begin{bmatrix} 1 & 1 \\ e^{2\pi jf} & e^{-2\pi jf} \\ e^{4\pi jf} & e^{-4\pi jf} \\ \vdots & \vdots \\ e^{2(N-1)\pi jf} & e^{-2(N-1)\pi jf} \end{bmatrix} \begin{bmatrix} \theta_1 \\ \theta_2 \end{bmatrix}$$

$$\underline{q}_1, \underline{q}_2 \in \mathbb{C}^N, \quad \underline{\theta} \in \mathbb{C}^2$$

Solution 2 Use $\cos(\alpha + \beta) = \cos(\alpha)\cos(\beta) - \sin(\alpha)\sin(\beta)$

$$s(n) = \underbrace{(D \cos(\phi))}_{\theta_1} \cos(2\pi fn) + \underbrace{(-D \sin(\phi))}_{\theta_2} \sin(2\pi fn)$$

$$\Rightarrow \underline{s} = \begin{bmatrix} 1 & 0 \\ \cos(2\pi f) & \sin(2\pi f) \\ \cos(4\pi f) & \sin(4\pi f) \\ \vdots & \vdots \\ \cos(2(N-1)\pi f) & \sin(2(N-1)\pi f) \end{bmatrix} \begin{bmatrix} \theta_1 \\ \theta_2 \end{bmatrix}$$

$$\underline{q}_1, \underline{q}_2 \in \mathbb{R}^N, \quad \underline{\theta} \in \mathbb{R}^2 \quad \text{if } D \in \mathbb{R}$$

Orthogonal Projection

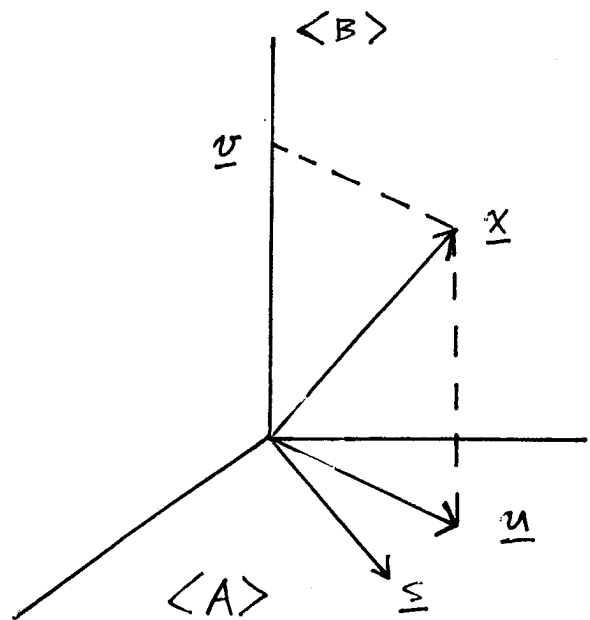
How can we use the knowledge of $\langle A \rangle$ to estimate \underline{s} from $\underline{x} = \underline{s} + \underline{w}$?

Since $\langle A \rangle$ and $\langle B \rangle$ are orthogonal complements, we can uniquely write

$$\underline{x} = \underline{u} + \underline{v}$$

where $\underline{u} \in \langle A \rangle$ and $\underline{v} \in \langle B \rangle$.

Since \underline{v} is pure noise, it makes sense to remove it.



In general, $\underline{s} \neq \underline{u}$ because \underline{w} has some component in $\langle A \rangle$.

Properties of projections

• $\pi_A^H =$ "self-adjoint"

• $\pi_A^2 =$ "idempotent"

(d)

• $\pi_A + \pi_B =$

• $\pi_A \cdot \pi_B =$

• If a_1, \dots, a_p are orthonormal, then

$$\pi_A = AA^H$$

Filtering interpretation

The projection operator is analogous to a bandpass filter; we only retain that information which resides in the passband, which corresponds to the signal subspace.

Least Squares Estimation

To estimate $\underline{s} = A\underline{\theta}$ where

$$\underline{x} = \underline{s} + \underline{w}$$

we use the projection onto $\langle A \rangle$:

$$\hat{\underline{s}} = \Pi_A \underline{x}$$

$$= A(A^H A)^{-1} A^H \underline{x}.$$

What if we want to estimate $\underline{\theta}$?

An estimate $\hat{\underline{\theta}}$ of $\underline{\theta}$ should satisfy

$$\hat{\underline{s}} = A\hat{\underline{\theta}}.$$

Therefore, an obvious estimate is

$$\hat{\underline{\theta}} =$$

②

It turns out that this is the solution to the least squares problem.

Proposition | The unique solution of

$$\min_{\underline{\theta}} \|\underline{x} - A\underline{\theta}\|^2 \quad (\underline{\theta} \in \mathbb{R}^p \text{ or } \mathbb{C}^p)$$

is $\hat{\underline{\theta}} = (A^H A)^{-1} A^H \underline{x}$.

Proof | Write $\underline{x} = \underline{u} + \underline{v}$ where $\underline{u} \in \langle A \rangle$
and $\underline{v} \in \langle A \rangle^\perp$. Observe

$$\begin{aligned} \|\underline{x} - A\underline{\theta}\|^2 &= \|\underline{u} - A\underline{\theta} + \underline{v}\|^2 \\ &= \langle \underline{u} - A\underline{\theta} + \underline{v}, \underline{u} - A\underline{\theta} + \underline{v} \rangle \\ &= \langle \underline{u} - A\underline{\theta}, \underline{u} - A\underline{\theta} \rangle + \langle \underline{v}, \underline{v} \rangle \\ &\quad + \underbrace{\langle \underline{u} - A\underline{\theta}, \underline{v} \rangle}_{=0} + \underbrace{\langle \underline{v}, \underline{u} - A\underline{\theta} \rangle}_{=0} \\ &= \|\underline{u} - A\underline{\theta}\|^2 + \|\underline{v}\|^2. \end{aligned}$$

The second term is independent of $\underline{\theta}$. Therefore, to minimize the expression, the best we can do is to make the first term 0 by taking

$$\underline{\theta} = \hat{\underline{\theta}} = (A^H A)^{-1} A^H \underline{x}.$$

Then $A\hat{\underline{\theta}} = \Pi_A \underline{x} = \underline{u}$. To see that $\hat{\underline{\theta}}$ is unique, if $\underline{\theta}'$ is also such that $\|\underline{u} - A\underline{\theta}'\| = 0$, then

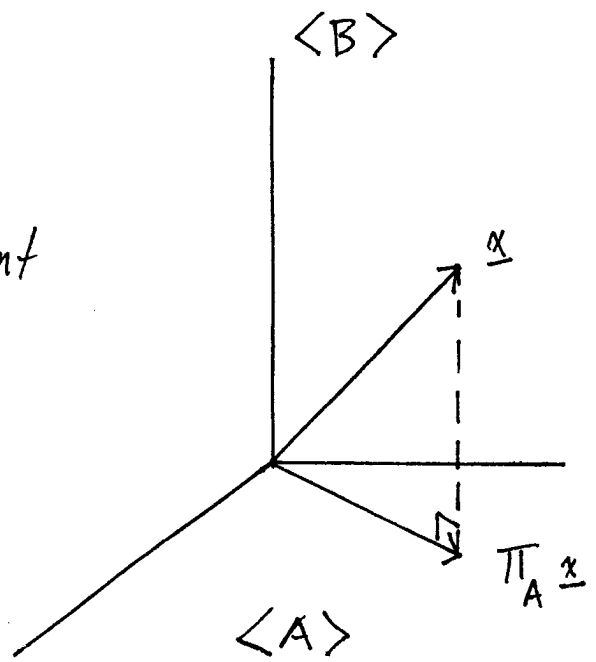
$$A\underline{\theta}' = \underline{u} \Rightarrow A\underline{\theta}' = A\hat{\underline{\theta}}$$

$$\Rightarrow \underline{\theta}' = \hat{\underline{\theta}}$$

since the columns of A are linearly independent. \square

Minimum distance property

We may conclude that $\Pi_A \underline{x}$ is the unique point in $\langle A \rangle$ that is closest to \underline{x} .



Remark | The operator

$$A^\# := (A^H A)^{-1} A^H$$

is called the pseudo-inverse of A .

Example Recall the sinusoid

$$s(n) = D \cdot \cos(2\pi f n + \phi) \quad , \quad n = 0, \dots, N-1$$

where f is known and D, ϕ are unknown.

How can we estimate D and ϕ from \underline{x} ?

Complex solution: $\underline{z} = A \underline{\theta}$ where

$$\underline{\theta} = \begin{bmatrix} \frac{D}{2} e^{j\phi} \\ \frac{D}{2} e^{-j\phi} \end{bmatrix}. \quad \text{Use pseudo-inverse to}$$

compute $\hat{\underline{\theta}}$ and form

$$\hat{D} = \sqrt{4 \cdot \hat{\theta}_1 \cdot \hat{\theta}_2} \quad , \quad \hat{\phi} = \frac{1}{2} \angle \left(\frac{\hat{\theta}_1}{\hat{\theta}_2} \right)$$

Real solution: $\underline{z} = A \underline{\theta}$ where

$$\underline{\theta} = \begin{bmatrix} D \cos(\phi) \\ -D \sin(\phi) \end{bmatrix}. \quad \text{Use pseudo-inverse to}$$

compute $\hat{\underline{\theta}}$ and form

$$\hat{D} = \sqrt{\hat{\theta}_1^2 + \hat{\theta}_2^2} \quad , \quad \hat{\phi} = \tan^{-1} \left(-\frac{\hat{\theta}_2}{\hat{\theta}_1} \right)$$

(assuming D to be real)

Summary

- If a signal lies in a subspace, it can be estimated by projection onto that subspace.
- This "filters out" any noise in the noise subspace.
- The projection satisfies the minimum distance property, and is closely related to the least squares problem.
- This approach is non-statistical because no probability model is specified for the noise. Yet it turns out to be equivalent or similar to many methods we will see later.

Key

- a. basis
- b. signal subspace, noise subspace
- c. orthogonal projection
- d. Π_A , Π_A , $I_{N \times N}$, $O_{N \times N}$
- e. $(A^H A)^{-1} A^H \mathbf{z}$