

Recitation 1 -EECS 451, Winter 2010

Jan 13, 2010

OUTLINE

- Review of Fourier Transform
- Practice problems

Concepts

1. Fourier Transforms

- Definition
- Basic Properties
- Conjugate symmetry

Problems

1. Determine the Fourier transform of the signal $x(t) = \frac{1}{\sqrt{2\pi}} e^{-t^2/2}$.
 - (a) Do you observe anything peculiar about the Fourier transform of this signal?
 - (b) Which other signal exhibits a similar property?
2. Compute the Fourier transform of each of the following signals
 - (a) $x(t) = e^{-|t|} \cos 2t$
 - (b) The signal $x(t)$ depicted in figure 1
3. Fig 2 shows an incomplete Fourier transform of a sinusoidal signal $x(t)$. Determine the signal $x(t)$ using the information given below.
 - (a) $x(t)$ is a real signal comprising of two nonzero frequencies.
 - (b) The value of the signal at time $t=0$ is 7.
 - (c) The energy of the signal is $(1+54\pi)/(2\pi)$

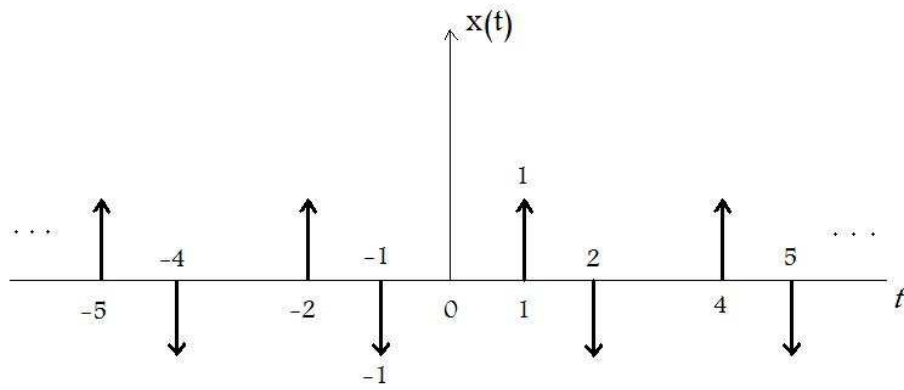


Figure 1:

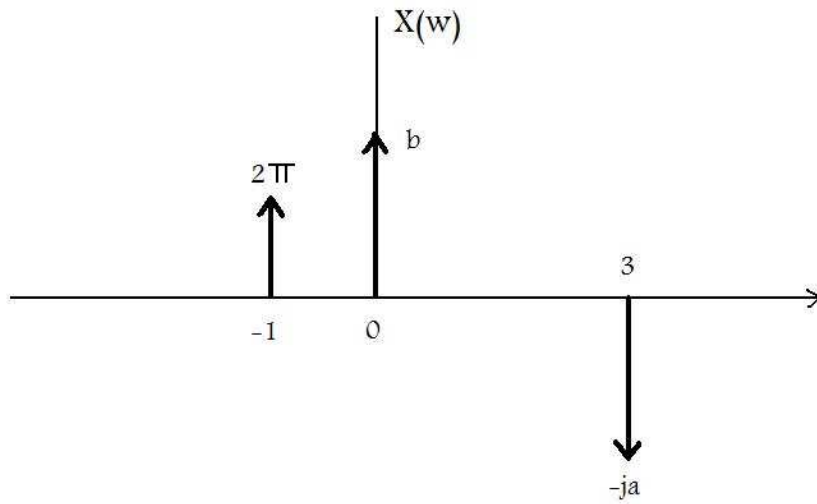


Figure 2:

$$1. \quad X(\omega) = \int_{-\infty}^{\infty} \frac{1}{\sqrt{2\pi}} e^{-\frac{t^2}{2}} e^{-j\omega t} dt$$

$$= \int_{-\infty}^{\infty} \frac{1}{\sqrt{2\pi}} e^{-\frac{1}{2}(t+j\omega)^2} e^{-\frac{1}{2}\omega^2} dt$$

(Let $t' = t + j\omega$)

$$= e^{-\frac{1}{2}\omega^2} \int_{-\infty}^{\infty} \frac{1}{\sqrt{2\pi}} e^{-\frac{t'^2}{2}} dt'$$

Since $\frac{1}{\sqrt{2\pi}} e^{-\frac{t'^2}{2}}$ is a pdf of a Gaussian R.V.,

$$\int_{-\infty}^{\infty} \frac{1}{\sqrt{2\pi}} e^{-\frac{t'^2}{2}} dt' = 1,$$

$$\therefore X(\omega) = e^{-\frac{1}{2}\omega^2}.$$

We can also prove $\int_{-\infty}^{\infty} \frac{1}{\sqrt{2\pi}} e^{-\frac{t^2}{2}} dt = 1$ as follows.

Since $\frac{1}{\sqrt{2\pi}} e^{-\frac{t^2}{2}}$ is non-negative for all t , it suffices to show that

$$\left[\int_{-\infty}^{\infty} \frac{1}{\sqrt{2\pi}} e^{-\frac{t^2}{2}} dt \right] \left[\int_{-\infty}^{\infty} \frac{1}{\sqrt{2\pi}} e^{-\frac{u^2}{2}} du \right] = 1.$$

$$\left[\int_{-\infty}^{\infty} \frac{1}{\sqrt{2\pi}} e^{-\frac{t^2}{2}} dt \right] \left[\int_{-\infty}^{\infty} \frac{1}{\sqrt{2\pi}} e^{-\frac{u^2}{2}} du \right]$$

$$= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} \frac{1}{2\pi} e^{-\frac{t^2+u^2}{2}} dt du.$$

(Let $t = r \cos \theta$, $u = r \sin \theta$, i.e. change the coordinate)

$$= \int_0^{2\pi} \int_0^{\infty} \frac{1}{2\pi} r e^{-\frac{r^2}{2}} dr d\theta$$

$$= \int_0^{2\pi} \frac{1}{2\pi} \left[-e^{-\frac{r^2}{2}} \right]_0^{\infty} d\theta$$

$$= \int_0^{2\pi} \frac{1}{2\pi} d\theta = 1.$$

(a) The original signal and the transform of it have the same form.

In other words, $x(t)$ is an eigen function of FT.

(b) Eigen functions of FT are called 'Hermite functions'.

One of them is $t e^{-\frac{t^2}{2}}$.

$$2. (a) \quad x(t) = e^{-|t|} \cos 2t$$

$$= e^{-|t|} \frac{e^{j2t} + e^{-j2t}}{2}$$

Let $Y(\omega)$ be the FT of $e^{-|t|}$.

Then, $X(\omega) = \frac{1}{2} (Y(\omega-2) + Y(\omega+2))$ (\because modulation property)

$$Y(\omega) = \int_{-\infty}^{\infty} e^{-|t|} e^{-j\omega t} dt$$

$$= \int_0^{\infty} e^{-t} e^{-j\omega t} dt + \int_{-\infty}^0 e^t e^{-j\omega t} dt$$

$$= \frac{1}{1+j\omega} + \frac{1}{1-j\omega} = \frac{2}{1+\omega^2}$$

$$\therefore X(\omega) = \frac{1}{1+(\omega-2)^2} + \frac{1}{1+(\omega+2)^2}$$

$$(b) \quad \text{Let } x_1(t) = \sum_{k=-\infty}^{\infty} \delta(t-3k).$$

Then, $x(t) = x_1(t-1) - x_1(t-2)$.

$$\text{Hence, } X(\omega) = e^{-j\omega} X_1(\omega) - e^{-j2\omega} X_1(\omega)$$

$$= (e^{-j\omega} - e^{-j2\omega}) X_1(\omega).$$

Consider now the Fourier series representation of $x_1(t)$.

$$\text{Then, } x_1(t) = \frac{1}{3} \sum_{k=-\infty}^{\infty} e^{j\frac{2\pi}{3}kt}$$

Since $1 \xrightarrow{\mathcal{F}} 2\pi \delta(\omega)$ and by the modulation property,

$$X_1(\omega) = \frac{2\pi}{3} \sum_{k=-\infty}^{\infty} \delta(\omega - \frac{2\pi}{3}k).$$

$$\therefore X(\omega) = \frac{2\pi}{3} (e^{-j\omega} - e^{-j2\omega}) \sum_{k=-\infty}^{\infty} \delta(\omega - \frac{2\pi}{3}k).$$

★ Please change (c) to 'The energy of the signal is $\frac{1+54\pi}{2\pi}$ '.

3. Since $x(t)$ has two nonzero frequencies, we know that there is no other frequency components than $0, \pm 1, \pm 3$.

Since $x(t)$ is real, $X(-\omega) = X^*(\omega)$ by conjugate symmetry.

Therefore, $X(-3) = ja$, $X(1) = 2\pi$.

$$\text{From (b), } X(0) = \frac{1}{2\pi} \int_{-\infty}^{\infty} x(\omega) d\omega = 7.$$

$$\frac{1}{2\pi} \int_{-\infty}^{\infty} x(\omega) d\omega = \frac{1}{2\pi} (4\pi + b)$$

$$\therefore b = 10\pi.$$

Fourier series coefficient of $x(t)$.

$$\text{From (c), } \frac{1}{T_0} \int_{T_0} |x(t)|^2 dt \stackrel{(i)}{=} \sum_k |a_k|^2 = \frac{1+54\pi}{2\pi} = \frac{1}{2\pi} + 27$$

where (i) is from Parseval's theorem for periodic signal,

and T_0 is the period of $x(t)$. Note that $x(t)$ is a periodic function. We can see that from the fact that the FT of $x(t)$ is discrete.

$$\text{Note that } X(\omega) = ja\delta(\omega+3) + 2\pi\delta(\omega+1) + 10\pi\delta(\omega) + 2\pi\delta(\omega-1) - ja\delta(\omega-3)$$

Then, by the definitions of FT and FS.

$$a_k \text{'s are } \frac{ja}{2\pi}, 1, 5, 1, -\frac{ja}{2\pi}.$$

$$\text{Therefore, } \sum_k |a_k|^2 = \frac{a^2}{2\pi^2} + 27 = \frac{1}{2\pi} + 27.$$

$$\therefore a = \pm\sqrt{\pi}.$$

$$\text{Then, } x(t) = \frac{5}{\sqrt{\pi}} + 2\cos t \pm \frac{1}{\sqrt{\pi}} \sin 2t.$$

We can prove that

$$\frac{1}{T_0} \int_{T_0} |x(t)|^2 dt = \sum_k |a_k|^2 \quad \text{as follows.}$$

$$x(t) = \sum_k a_k e^{j\frac{2\pi}{T_0}kt}, \quad \text{where } a_k = \frac{1}{T_0} \int_{T_0} x(t) e^{-j\frac{2\pi}{T_0}kt} dt.$$

$$\text{Therefore, } \frac{1}{T_0} \int_{T_0} x^*(t) e^{-j\frac{2\pi}{T_0}kt} dt = a_{-k}^*,$$

$$\text{and hence, } x^*(t) = \sum_k a_{-k}^* e^{j\frac{2\pi}{T_0}kt}.$$

$$\text{Let } x(t)x^*(t) = \sum_k c_k e^{j\frac{2\pi}{T_0}kt}$$

$$\text{Then, } c_k = \frac{1}{T_0} \int_{T_0} x(t)x^*(t) e^{-j\frac{2\pi}{T_0}kt} dt$$

$$= \frac{1}{T_0} \int_{T_0} \left[\sum_l a_l e^{j\frac{2\pi}{T_0}lt} \right] \left[\sum_m a_m^* e^{j\frac{2\pi}{T_0}mt} \right] e^{-j\frac{2\pi}{T_0}kt} dt$$

$$= \frac{1}{T_0} \int_{T_0} \sum_l \sum_m a_l a_m^* e^{j\frac{2\pi}{T_0}t(l+m-k)} dt.$$

$$= \sum_l \sum_m a_l a_m^* \int_{T_0} \frac{1}{T_0} e^{j\frac{2\pi}{T_0}t(l+m-k)} dt.$$

$$\left(\text{Since } \frac{1}{T_0} \int_{T_0} e^{j\frac{2\pi}{T_0}kt} dt = \begin{cases} 1 & k=0 \\ 0 & \text{otherwise} \end{cases} \right)$$

$$= \sum_l a_l a_{l-k}^*$$

$$\text{Then, } \frac{1}{T_0} \int_{T_0} |x(t)|^2 dt$$

$$= \frac{1}{T_0} \int_{T_0} x(t)x^*(t) dt$$

$$= \frac{1}{T_0} \int_{T_0} \sum_k c_k e^{j\frac{2\pi}{T_0}kt} dt$$

$$= c_0$$

$$= \sum_l a_l a_l^* = \sum_l |a_l|^2.$$

We can prove the similar thing for non-periodic signal as follows.

$$\begin{aligned} & \int_{-\infty}^{\infty} |x(t)|^2 dt \\ &= \int_{-\infty}^{\infty} x(t) x^*(t) dt \\ &= \int_{-\infty}^{\infty} x(t) \left[\frac{1}{2\pi} \int_{-\infty}^{\infty} x^*(\omega) e^{j\omega t} d\omega \right] dt \quad (\because x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} x(\omega) e^{j\omega t} d\omega) \\ &= \frac{1}{2\pi} \int_{-\infty}^{\infty} x^*(\omega) \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt d\omega \quad (\because \text{change the order of integration}) \\ &= \frac{1}{2\pi} \int_{-\infty}^{\infty} x^*(\omega) X(\omega) d\omega \\ &= \frac{1}{2\pi} \int_{-\infty}^{\infty} |X(\omega)|^2 d\omega. \end{aligned}$$