

Given: 2^{nd} -order Butterworth filter: $H_a(s) = \frac{1}{s^2 + \sqrt{2}s + 1}$.

Goal: Use **bilinear transform** to design a digital lowpass filter with cutoff frequency $\omega_o = \frac{\pi}{2}$.

Sol'n: $1 = \Omega_o = \frac{2}{T} \tan(\frac{\omega_o T}{2}) = \frac{2}{T} \tan(\frac{\pi/2}{2}) \rightarrow T = 2$. $s = \frac{2}{T} \frac{z-1}{z+1} = \frac{z-1}{z+1}$.

Filter: Substitute in $H_a(s)$: $H(z) = \frac{1}{[\frac{z-1}{z+1}]^2 + \sqrt{2}[\frac{z-1}{z+1}] + 1} = \frac{(z+1)^2}{(z+1)^2 + \sqrt{2}(z+1) + 1}$.

Simplify: $H(z) = \frac{(z+1)^2}{(z-1)^2 + \sqrt{2}(z-1)(z+1) + (z+1)^2} = \frac{(z+1)^2}{(2+\sqrt{2})z^2 + (2-\sqrt{2})}$.

Zeros: $\{-1, -1\} \rightarrow$ rejects high freqs. **Poles:** $\{\pm j\sqrt{\frac{2-\sqrt{2}}{2+\sqrt{2}}}\}$.

Goal: Use **frequency sampling** at $\omega = \frac{\pi}{3}, \frac{2\pi}{3}$ to design a 5-point digital differentiator.

Sol'n: $H(e^{j\omega}) = j\omega$, $|\omega| < \pi$ is imaginary and odd.

Form: $h[n]$ should be odd. Let $h[n] = \{b, a, 0, -a, -b\}$.

$H(e^{j\omega}) = be^{j2\omega} + ae^{j\omega} - ae^{-j\omega} - be^{-j2\omega} = j2b \sin(2\omega) + j2a \sin(\omega)$.

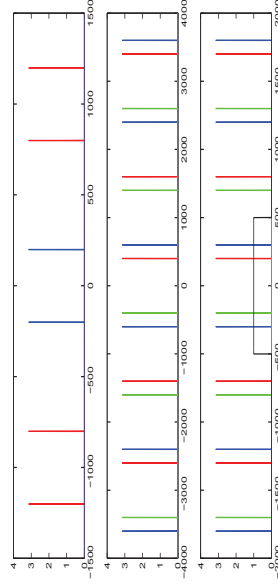
$\omega = \frac{\pi}{3}$: $H(e^{j\pi/3}) = j\frac{\pi}{3} = j2a\frac{\sqrt{3}}{2} + j2b\frac{\sqrt{3}}{2} \rightarrow a+b = \frac{\pi}{3\sqrt{3}}$.

$\omega = \frac{2\pi}{3}$: $H(e^{j2\pi/3}) = j\frac{2\pi}{3} = j2a\frac{\sqrt{3}}{2} - j2b\frac{\sqrt{3}}{2} \rightarrow a-b = \frac{2\pi}{3\sqrt{3}}$.

Solve: $a = \frac{\pi}{2\sqrt{3}} \approx 0.9$. $b = -\frac{\pi}{6\sqrt{3}} \approx -0.3$ (rounding slightly).

Answer: $h[n] = \{-0.3, 0, 0, -0.9, 0.3\}$. Compare to previous slide.

Given: $\cos(2\pi \frac{200}{1000} n) \rightarrow \downarrow 3 \rightarrow z[n] \rightarrow \uparrow 2 \rightarrow \boxed{\text{LPF}} \rightarrow y[n]$.



Goal: Use a rectangular **window** to design a 5-point (length=5) digital differentiator.

Sol'n: $h_{\text{ideal}}[n] = \text{DTFT}^{-1}\{j\omega\} = \frac{1}{2\pi} \int_{-\pi}^{\pi} (j\omega) e^{j\omega n} d\omega = \frac{(-1)^n}{n}$.

Window: $w[n] = \{1, 1, 1, 1, 1\}$. Then $h[n] = h_{\text{ideal}}[n]w[n]$: $(h[0]=0)$

Answer: $h[n] = \{-\frac{1}{2}, 1, 0, -1, \frac{1}{2}\}$. Compare to the next slide.

Filter: $y[n] = (x[n+1] - x[n-1]) - \frac{1}{2}(x[n+2] - x[n-2])$. Differences!

Given: $\cos(2\pi \frac{200}{1000} n) \rightarrow \downarrow 3 \rightarrow z[n] \rightarrow \uparrow 2 \rightarrow y[n]$.

Goal: Plot spectra of $x[n]$, $z[n]$, $y[n]$. Use $f = \frac{1000\omega}{2\pi}$ Hz.

And: Show output has components at 200 Hz and 300 Hz.

Sol'n: See 3 plots on next slide. All spectra periodic at 1000 Hz.

Given: $\cos(2\pi \frac{200}{1000} n) \rightarrow \uparrow 2 \rightarrow z[n] \rightarrow \boxed{\text{LPF}} \rightarrow \downarrow 3 \rightarrow y[n]$.

Goal: Plot spectra of $x[n]$, $z[n]$, $y[n]$. Use $f = \frac{1000\omega}{2\pi}$ Hz.

And: Show output has components only at 300 Hz.

Sol'n: See 3 plots on next slide. All spectra periodic at 1000 Hz.

Given: $\cos(2\pi \frac{200}{1000} n) \rightarrow \uparrow 2 \rightarrow z[n] \rightarrow \boxed{\text{LPF}} \rightarrow \downarrow 3 \rightarrow y[n]$.

