## FORWARD OR DIRECT Z-TRANSFORM

| Huh?              | $\begin{aligned} \mathcal{Z}\{x[n]\} &= X(z) = \sum_{n=0}^{\infty} x[n] z^{-n} \text{ (compare to } \mathcal{L}\{x(t)\} = \int_{0}^{\infty} x(t) e^{-st} dt \text{).} \\ \text{Finite-duration signal } x[n] \to \text{polynomial } X(z) \text{ with coefficients} = x[n]. \\ \text{Infinite-duration } x[n] = \text{coefficients of Laurent (power) series } X(z). \end{aligned}$ |
|-------------------|--|
| $\mathbf{length}$ | $\begin{aligned} x[n] &= \{\underline{3}, 1, 4, 2, 5\}. \text{ Note } x[0] = 3 \text{ and finite duration} = 5 \rightarrow \\ X(z) &= 3 + 1z^{-1} + 4z^{-2} + 2z^{-3} + 5z^{-4} = (3z^4 + z^3 + 4z^2 + 2z + 5)/z^4. \\ \mathcal{Z}\{\delta[n-D]\} &= z^{-D} \text{ for any integer delay } D \ge 0. \ \mathcal{Z}\{\delta[n]\} = 1. \end{aligned}$                                 |
|                   | $\begin{aligned} \mathcal{Z}\{a^{n}u[n]\} &= \sum_{n=0}^{\infty} a^{n}z^{-n} = \sum_{n=0}^{\infty} (az^{-1})^{n} = 1/(1-az^{-1}) = z/(z-a).\\ \text{EX: } \mathcal{Z}\{(\frac{1}{2})^{n}u[n]\} &= \frac{1}{1-\frac{1}{2}z^{-1}}. \text{ EX: } \mathcal{Z}\{(-\frac{1}{3})^{n}u[n]\} = \frac{1}{1+\frac{1}{3}z^{-1}}. \end{aligned}$  |
|                   | $\mathcal{Z}\{\cos(\omega_0 n)u[n]\} = \frac{1}{2}\mathcal{Z}\{e^{j\omega_0 n}u[n]\} + \frac{1}{2}\mathcal{Z}\{e^{-j\omega_0 n}u[n]\} = (\text{since linear})$ $\frac{1}{2}\frac{1}{1-e^{j\omega_0 z^{-1}}} + \frac{1}{2}\frac{1}{1-e^{-j\omega_0 z^{-1}}} = \frac{1-z^{-1}\cos(\omega_0)}{1-2z^{-1}\cos(\omega_0)+z^{-2}} = \frac{z^2-z\cos(\omega_0)}{z^2-2z\cos(\omega_0)+1}.$  |
| Linear:           | $\mathcal{Z}\{\{\underline{3},1,4\}+u[n]\} = \frac{3z^2+1z+4}{z^2} + \frac{z}{z-1} = \frac{4z^3-2z^2+3z-4}{z^3-z^2} = \frac{\text{RATIONAL}}{\text{FUNCTION}}.$  |
| Delay:            | If $\mathcal{Z}{x[n]} = X(z)$ and $D \ge 0$ , then $\mathcal{Z}{x[n-D]} = z^{-D}X(z)$ .  |
| EX:               | $\mathcal{Z}\{u[n] - u[n-2]\} = \frac{z}{z-1} - z^{-2}\frac{z}{z-1} = \frac{z}{z^2}\frac{z^2-1}{z-1} = 1 + z^{-1}.$  |
| Note:             | Note that $u[n] - u[n-2] = \delta[n] + \delta[n-1]$ , so these are consistent.   |
| Also:             | $\mathcal{Z}\left\{\frac{1}{n}u[n-1]\right\} = \sum_{n=1}^{\infty} \frac{z^{-n}}{n} = -\log(1-z^{-1})$ if you recognize this.  |
|                   | Convolution $\leftrightarrow$ polynomial multiplication: $\mathcal{Z}\{x[n] * y[n]\} = X(z)Y(z).$  |
| Eigen-            | $z^n \to \overline{ h[n] } \to z^n H(z)$ : $z^n$ in $\to$ scaled $z^n$ out. $H(z) = \sum_{n=0}^{\infty} h[n] z^{-n}$ .   |
|                   | $h[n] * z^{\overline{n}} = \sum h[i] z^{n-i} = z^n \sum h[i] z^{-i} = z^n H(z). \ H(e^{j\omega}) = H(z) _{z=e^{j\omega}}.$   |
| of LTI            | Note $z^n$ plays the same role that $e^{st}$ plays in continuous time.   |
| Note:             | $H(z) = $ transfer function. $H(e^{j\omega}) = H(z) _{z=e^{j\omega}} = $ frequency response.   |
| EX:               | Compute step response (to $u[n]$ ) of LTI system with $h[n] = \{\underline{2}, -3, 1\}$ .  |
| Soln:             | We need to compute $y[n] = \{\underline{2}, -3, 1\} * u[n]$ . Do this two ways:  |
| #1:               | $y[n] = (2\delta[n] - 3\delta[n-1] + 1\delta[n-2]) * u[n]$ . Using $u[n] * \delta[n] = u[n]$ ,   |
| #1:               | $y[n] = 2u[n] - 3u[n-1] + u[n-2] = \{\underline{2}, -1\}$ (try it!) has duration=2.  |
| #2:               | $Y(z) = H(z)U(z) = (2 - 3z^{-1} + z^{-2})\frac{1}{1 - z^{-1}} = 2 - z^{-1} \to y[n] = \{\underline{2}, -1\}.$  |
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Note: This system has wiped out the step input! This seldom happens.

## APPLICATIONS OF THE z-TRANSFORM

| Given:                     | $x[n] = 3^{n}u[n] \to \overline{ y[n] - 2y[n-1]} = x[n-1] - x[n-2]  \to y[n]$   |
|----------------------------|---|
| Goal:                      | Compute the response $y[n]$ of the system to this particular input.   |
| $\mathcal{Z}$ :            | $Y(z) - 2z^{-1}Y(z) = z^{-1}X(z) - z^{-2}X(z)$ and $X(z) = \frac{z}{z-3}$ here  |
| $\rightarrow$              | $Y(z) = \frac{z^{-1} - z^{-2}}{1 - 2z^{-1}} \frac{z}{z - 3} = \frac{z - 1}{(z - 2)(z - 3)} = \frac{2}{z - 3} - \frac{1}{z - 2}  \left[2 = \frac{3 - 1}{3 - 2}; -1 = \frac{2 - 1}{2 - 3}\right]$ |
| $\rightarrow$              | $y[n] = [2(3)^{n-1} - (2)^{n-1}]u[n-1] = \operatorname{FORCED}_{\operatorname{RESPONSE}}(\operatorname{like}_{x[n]}) + \operatorname{RESPONSE}(\operatorname{like}_{h[n]}).$                    |
| Given:                     | $x[n] = (\frac{1}{2})^n u[n] \to \overline{ \mathbf{LTI} } \to y[n] = \{\underline{0}, 0, 1\} = \delta[n-2]$  |
| Goal:                      | Compute the response of this system to $2\cos(\frac{\pi}{3}n)$ .  |
| H(z):                      | $\frac{\text{TRANSFER}}{\text{FUNCTION}} = H(z) = \mathcal{Z}\{y[n]\} / \mathcal{Z}\{x[n]\} = z^{-2} / [z/(z-\frac{1}{2})] = (z-\frac{1}{2})/z^3.$  |
| h[n]:                      | $ {}^{\rm IMPULSE}_{\rm RESPONSE} = h[n] = \mathcal{Z}^{-1}\{(z - \frac{1}{2})/z^3\} = \mathcal{Z}^{-1}\{z^{-2} - \frac{1}{2}z^{-3}\} = \{\underline{0}, 0, 1, -\frac{1}{2}\}. $                |
| H(w):                      | $\underset{\text{RESPONSE}}{\text{FREQUENCY}} = H(\omega) = H(z) _{z=e^{j\omega}} = (e^{j\omega} - \frac{1}{2})/e^{j3\omega}.$  |
| $\omega = \frac{\pi}{3}$ : | $H(\frac{\pi}{3}) = [e^{j\pi/3} - \frac{1}{2}]/e^{j3\pi/3} = (\frac{1}{2} + j\frac{\sqrt{3}}{2} - \frac{1}{2})/(-1) = -j\frac{\sqrt{3}}{2} = \frac{\sqrt{3}}{2}e^{-j\pi/2}.$                    |
| Sol'n:                     | $2\cos(\frac{\pi}{3}n) \to \overline{ \mathbf{LTI} } \to \sqrt{3}\cos(\frac{\pi}{3}n - \frac{\pi}{2}) = \sqrt{3}\sin(\frac{\pi}{3}n).$  |
| Given:                     | $x[n] \to \overline{ y[n] = x[n] - \frac{3}{4}x[n-1] + \frac{1}{8}x[n-2] } \to y[n]$  |
| Huh?                       | x[n]=cell phone signal. $y[n]$ =multipath due to buildings.   |
|                            | Compute the <b>inverse filter</b> that recovers $x[n]$ from $y[n]$ :  |
| Huh?                       | $x[n] \to \underline{ h[n] } \to y[n] \to \underline{ g[n] } \to x[n]$ . That is, $g[n]$ undoes $h[n]$ .  |
| Idea:                      | Systems in cascade (series) $\Leftrightarrow h[n] * g[n] = \delta[n] \Leftrightarrow H(z)G(z) = 1.$   |
| Here:                      | $h[n] = \{\underline{1}, -\frac{3}{4}, \frac{1}{8}\} \to H(z) = 1 - \frac{3}{4}z^{-1} + \frac{1}{8}z^{-2} = (z^2 - \frac{3}{4}z + \frac{1}{8})/z^2.$  |
| $\rightarrow$              | $G(z) = 1/H(z) = \frac{z^2}{[z^2 - \frac{3}{4}z + \frac{1}{8}]} = \frac{z^2}{[(z - \frac{1}{2})(z - \frac{1}{4})]}.$  |
| $\mathcal{Z}^{-1}$ :       | $\frac{G(z)}{z} = \frac{z}{(z-1/2)(z-1/4)} = \frac{2}{z-1/2} - \frac{1}{z-1/4} \to G(z) = 2\frac{z}{z-1/2} - 1\frac{z}{z-1/4}.$   |
| using:                     | residues $2 = (1/2)/[(1/2) - (1/4)]$ and $-1 = (1/4)/[(1/4) - (1/2)]$ .   |
| $\mathbf{g}[\mathbf{n}]$ : | $g[n] = 2(\frac{1}{2})^n u[n] - (\frac{1}{4})^n u[n]$ =inverse filter for original system.  |
| Note:                      | Stable since zeros of $H(z)$ =poles of $G(z)$ are inside unit circle.   |
| Note:                      | $g[0] \neq 0$ since both $\underset{\text{denominator}}{\overset{\text{numerator }\&}{\text{denominator}}}$ of $G(z)$ have the same degrees.  |